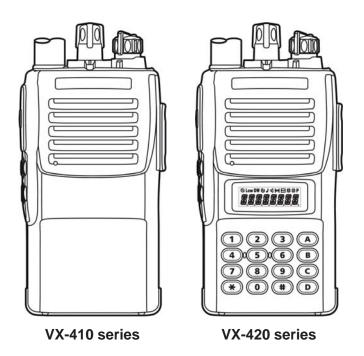


VHF FM Transceiver VX-410/-420 series Service Manual

Vertex Standard LMR, Inc.

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Introduction

This manual provides technical information necessary for servicing the VX-410/-420 series VHF FM Transceiver.

Servicing this equipment requires expertise in handling surface-mount chip components. Attempts by non-qualified persons to service this equipment may result in permanent damage not covered by the warranty, and may be illegal in some countries.

Two PCB layout diagrams are provided for each doublesided circuit board in the transceiver. Each side of is referred to by the type of the majority of components installed on that side ("leaded" or "chip-only"). In most cases one side has only chip components, and the other has either a mixture of both chip and leaded components (trimmers, coils, electrolytic capacitors, ICs, etc.), or leaded components only.

While we believe the technical information in this manual to be correct, VERTEX STANDARD assumes no liability for damage that may occur as a result of typographical or other errors that may be present. Your cooperation in pointing out any inconsistencies in the technical information would be appreciated.

Important Note

After Lot.36 of this transceiver was assembled using Pb (lead) free solder, based on the RoHS specification.

Only lead-free solder (Alloy Composition: Sn-3.0Ag-0.5Cu) should be used for repairs performed on this apparatus. The solder stated above utilizes the alloy composition required for compliance with the lead-free specification, and any solder with the above alloy composition may be used.

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Specifications

GENERAL Specifications

- · · · · · · · · · · · · · ·	
Frequency Range:	146 - 174 MHz
Number of Channels:	32 (16+16) channels
Number of Groups:	2 groups
Channel Spacing:	12.5 / 20 / 25 kHz
PLL Steps	2.5 / 6.25 kHz
Operating Voltage:	7.5 VDC ± 20%
Battery Life (w/save):	12.5 h (16 h) w/ FNB-V67LI (Rx 5 : Tx 5 : STBY 90 Duty)
	7 h (8.8 h) w/ FNB-V57 (Rx 5 : Tx 5 : STBY 90 Duty)
Temperature Range:	–22 °F to +140 °F (–30 °C to +60 °C)
Frequency Stability:	±2.5 ppm
Dimensions (WxHxD):	2.3" x 4.3" x 1.2" (58 x 18.5 x 30 mm) w/FNB-V67LI
Weight (approx.):	10.6 oz (300 g) w/Antenna, Battery (FNB-V67LI), and Belt Clip

RECEIVER Specifications (Measurements made per TIA/EIA-603)

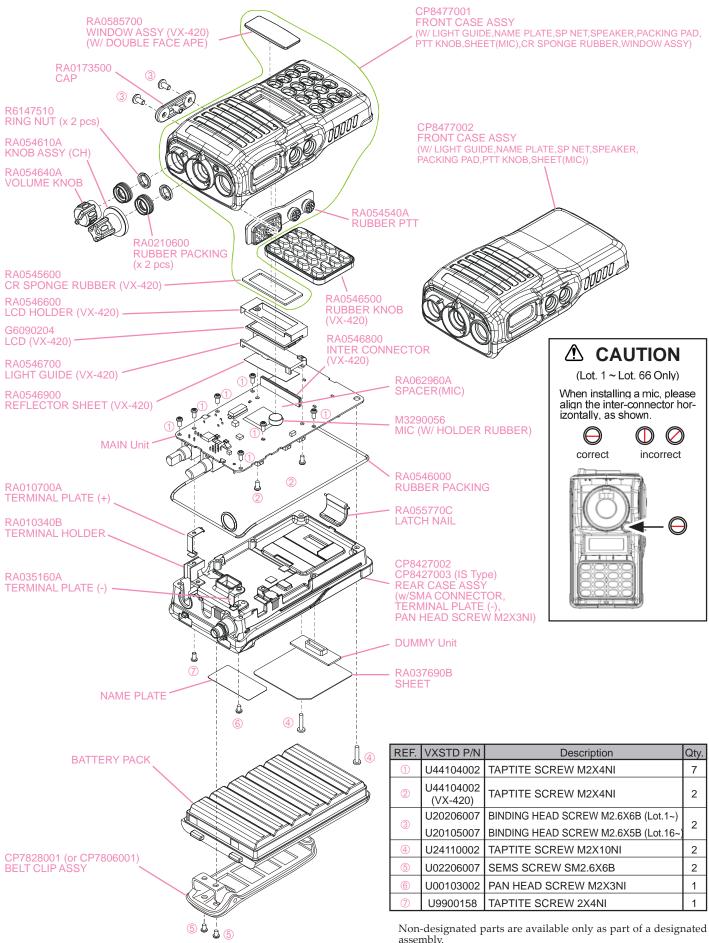
Sensitivity (12 dB SINAD):	0.25 μV
Adjacent Channel Selectivity (W/N):	70 / 65 dB
Intermodulation (W/N):	70 dB
Spurious and Image Rejection:	70 dB
Hum and Noise (W/N):	45 / 40 dB
Audio output:	500 mW @4 Ohms, 5% THD
-	

TRANSMITTER Specifications (Measurements made per TIA/EIA-603)

Power output:	5 / 1 W
Modulation (W/N):	16K0F3E, 11K0F3E
Spurious Emissions:	70 dB
FM Hum and Noise Analog (W/N):	45 / 40 dB
Audio Distortion:	Less than 3% @ 1 kHz

Measurements per EIA standards unless noted above. Specifications subject to change without notice or obligation.

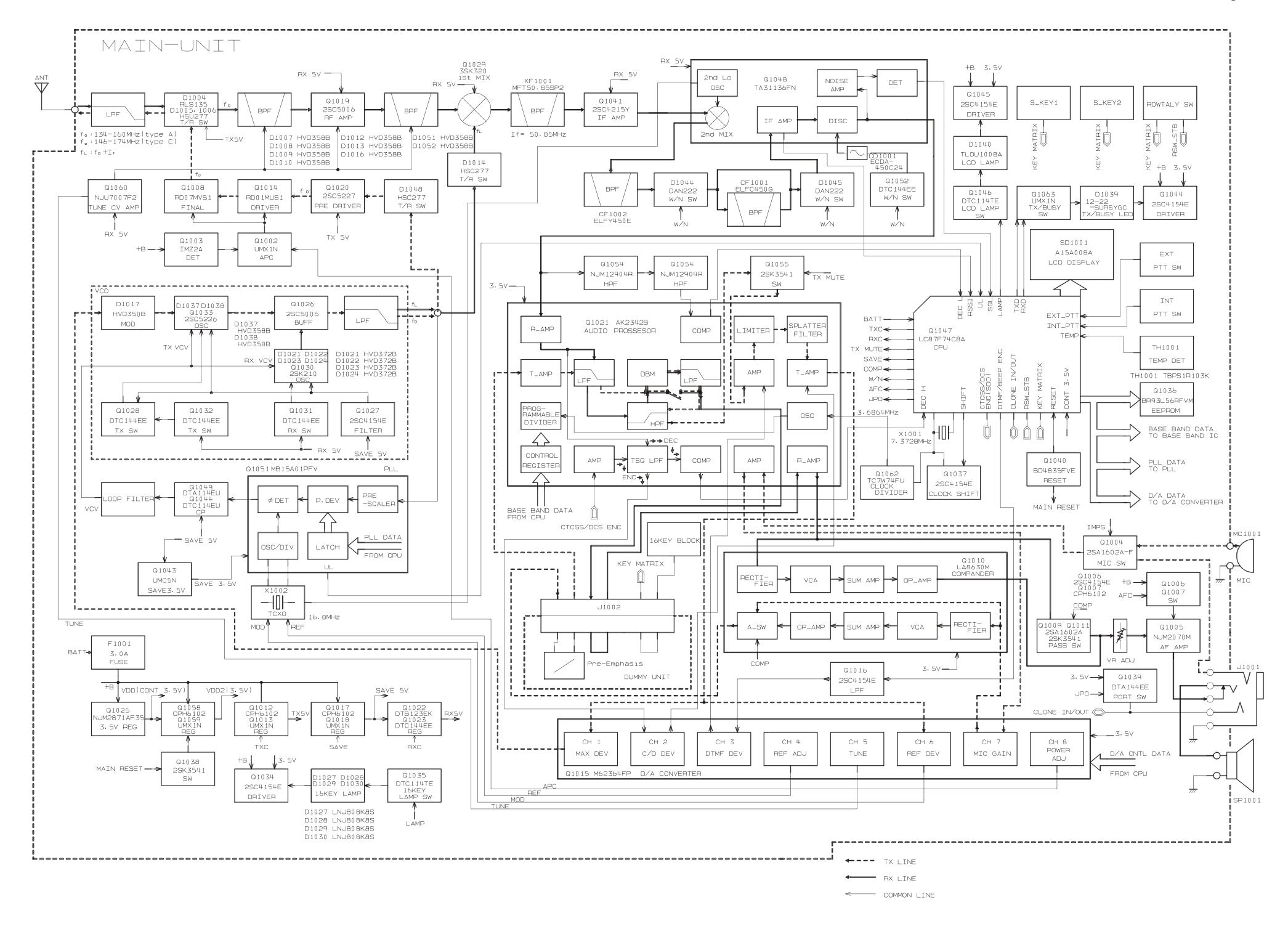
Exploded View & Miscellaneous Parts



VX-410/-420 VHF Series Service Manual

Parts List

REF.	DESCRIPTION	VALUE	V/W	TOL.	MFR'S DESIG	VXSTD P/N	VERS.	LOT.	SIDE	LAY ADR
	Main Unit PCB with Com	ponents				CB3575003	VX-420			
						CB3575004	VX-410			
						CB3575017	VX-420 (IS T	ype)		
							VX-410 (IS T	ype)		
	Dummy Unit PCB with Co	omponents				CB2731001				
	FRONT CASE ASSY					CP8477001				
L						CP8477002	VX-410			
	REAR CASE ASSY					CP8427002 CP8427003	IS Tupo			
	VOLUME KNOB					RA054640A	13 Туре			
	KNOB ASSY				(CH)	RA054610A				
	RING NUT				(01)	R6147510				
	RUBBER PACKING					RA0210600				
	RUBBER PTT					RA054540A				
	RUBBER KNOB					RA0546500	VX-420			
	LCD HOLDER					RA0546600				
	INTER CONNECTOR					RA0546800	VX-420			
I	PAN HEAD SCREW	1 pc			M2X3NI	U00103002				
	SEMS SCREW	2 pcs			SM2.6X5B	U02205007				
	BINDING HEAD SCREW				M2.6X6B	U20206007				
I	TAPTITE SCREW	2pcs 7 pcs			M2X10NI M2X4NI	U24110002 U44104002				
I	TAPTITE SCREW	7 pcs 2 pcs			M2X4NI M2X4NI	U44104002	VX-420			
I	TAPTITE SCREW	2 pcs 1 pc			2X4NI	U9900158	V TLU			
	SPEAKER				36N-D2114B 1.0W/4-OHM	M4090157				
CD1001	CERAMIC DISC				JTBM450CX24-A	H7901530A		1-	В	c2
CF1001	CERAMIC FILTER				LTUCG450G-A	H3900566A		1-	В	c3
	CERAMIC FILTER				LTWC450E-A	H3900565A		1-	В	c3
DS1001					GTO4966SY01	G6090204	VX-420	1-	Α	C2
F 1001	CHIP FUSE	3.15A			FHC16 322ADTP	Q0000118	10 T	1-	A	E3
F 1001	CHIP FUSE MICROPHONE ELEMENT	3A			0434 003. 3.0A 0B-22S44-C1033MG H/F	Q0000107	IS Type	1-	A	E3 D1
Q 1001	FET				0B-22S44-C1033MG H/F RD07MVS1A-T12	M3290056 G3070352		1-	A B	D1 b1
S 1008	ROTARY SWITCH				TP70TF5161	N0190183		1-	B	a2
S 1002 S 1003	TACT SWITCH				SKRTLAE010	N5090130		1-	В	a2 a1
S 1004	TACT SWITCH				SKRTLAE010	N5090130		1-	В	b1
S 1005	TACT SWITCH				SKRTLAE010	N5090130		1-	В	c1
	THERMISTOR				ERTJ1VR103J	G9090118		1-	А	D2
	THERMISTOR				ERTJ0EV473J	G9090120		1-	А	C2
VR1001	-				R087A0NSFX13.5C6.5A20300I			1-	В	a3
		7.3728MHz			7.3728MHZ	H0103280		1-	A	C3
X 1002	TCXO	16.8MHz			TTS05VS-M1 16.8MHZ	H9500830		1-	В	d2
XF1001	XTAL FILTER				MFT50.85P2 50.85MHZ	H1102364		1-	В	b2
-										



Block Diagram

Circuit Description

Receive Signal Path

Incoming RF from the antenna jack is delivered to the RF Unit and passes through a low-pass filter consisting of coils L1002, and L1003, capacitors C1001, C1002, C1021, C1024, C1025, C1026, and C1027, and antenna switching diode D1004 (**RLS135**).

Signals within the frequency range of the transceiver enter a varactor-tuned band-pass filter consisting of coils L1012 and L1015, capacitors C1058, C1059, C1085, C1118, and C1119, and diodes D1007, D1008, D1009, and D1010 (all **HVD350B**), then are amplified by Q1019 (**2SC5006**) and enter a varactor-tuned band-pass filter consisting of coils L1020, and L1025, capacitors C1053, C1054, C1056, C1180, and C1181, and diodes D1012, D1013, and D1016 (all **HVD350B**), before application to the first mixer, Q1029 (**3SK320**).

Buffered output from the VCO is amplified by Q1026 (**2SC5005**) to provide a pure first local signal between 196.85 and 224.85 MHz for injection to the first mixer Q1029. The 50.85 MHz first mixer product then passes through monolithic crystal filter XF1001 (**MFT50.85PT**, 5.5 kHz BW) to strip away unwanted mixer products, and is then amplified by Q1041 (**2SC4215Y**).

The amplified first IF signal is applied to FM IF subsystem IC Q1048 (**TA31136FN**), which contains the second mixer, second local oscillator, limiter amplifier, noise amplifier, and RSSI amplifier.

The second local signal is produced from the PLL reference/second local oscillator of X1002 (TCXO TTS05VS-M1 16.80 MHz). The 16.80 MHz reference signal is tripled by Q1048, capacitor C1251, and coil L1042, and the resulting the 50.4 MHz second local signal is then delivered to the mixer section of Q1048 which produces the 450 kHz second IF when mixed with the first IF signal.

The second IF then passes through the ceramic filter CF1001 (**ELFC450G** on "Narrow" channels) or CF1002 (**ELFY450E** on "Wide" and "Narrow" channels) to strip away all but the desired signal, and is then applied to the limiter amplifier in Q1048, which removes amplitude variations in the 450 kHz IF, before detection of the speech by the ceramic discriminator CD1001 (**ECDA450C24**).

Squelch Control

The squelch circuitry consists of a noise amplifier, band-pass filter, and noise detector within Q1048 (TA31136FN).

When no carrier is received, noise at the output of the detector stage in Q1048 is amplified and band-pass filtered by the noise amplifier section of Q1048 and the network between pins 7 and 8, and then is rectified by detection circuit in Q1048.

The resulting DC squelch control voltage is passed to pin 19 of the microprocessor Q1047 (**LC87F74C8A**). If no

carrier is received, this signal causes pin 19 of Q1047 to go high and pin 30 to go high. Pin 35 signals Q1006 (**2SC4154E**) to disable the supply voltage to the audio amplifier Q1005, while pin 30 holds the green (Busy) half of the LED off, when pin 35 is high and pin 30 is high.

Thus, the microprocessor blocks output from the audio amplifier, and silences the receiver, while no signal is being received (and during transmission, as well).

When a carrier appears at the discriminator, noise is removed from the output, causing pin 19 of Q1047 to go low and the microprocessor to activate the "Busy" LED via Q1047.

The microprocessor then checks for CTCSS or CDCSS code squelch information, if enabled. If not transmitting and CTCSS or CDCSS is not activated, or if the received tone or code matches that programmed, audio is allowed to pass through the audio amplifier Q1005 (**NJM2070M**) to the loudspeaker by the enabling of the supply voltage to it via Q1048.

Transmit Signal Path

Speech input from the microphone MC1001 passes through the audio amplifier Q1021 (**AK2342B**) to Q1015 (**M62364FP**) which adjusts the microphone gain. The adjusted audio is applied to the compander Q1010 (**LA8630M**) which compresses the speech signal according to a control command from the microprocessor Q1047 (**LC87F74C8A**).

The compressed speech signal passes through the dummy unit and pre-emphasis circuit to Q1021, which contains the low-pass filter, Voice Scrambler selector, and high-pass filter.

The output from Q1021 is applied to the AF mute gate Q1055 (**2SK3541**), then returns to Q1021, which contains the limiter amplifier, splatter filter and audio amplifier.

The filtered audio signal is applied to Q1015 (**M62364FP**) which is adjusts the audio level, then is applied to varactor diode D1017 (**HVD350B**), which frequency modulates the VCO Q1033 (**2SK508**). A portion of the audio signal from Q1015 is applied to TCXO X1002 (**TTS05VS**).

The processed audio may then be mixed with a CTCSS tone generated by Q1047 (**LC87F74C8A**) for frequency modulation of the PLL carrier (up to ±5 kHz from the unmodulated carrier) at the transmitting frequency.

If a CDCSS code is enabled for transmission, the code is generated by microprocessor Q1047 and delivered to X1002 (**TCXO TTS05VS**) for CDCSS modulating.

The modulated signal from the VCO Q1033 (**2SC5226**) is buffered by Q1026 (**2SC5005**). The low-level transmit signal then passes through the T/R switching diode D1048 (**HSC277TRF**) to the driver amplifier Q1014 (**2SK2596BXTL**), and then the amplified transmit signal

is applied to the final amplifier Q1008 (**RD07MVS1**), which delivers up to 5 watts of output power.

The transmit signal then passes through the antenna switch D1004 (**RLS135**) and is low-pass filtered, to suppress harmonic spurious radiation before delivery to the antenna.

Automatic Transmit Power Control

Current from the final amplifier is sampled by R1015, R1038 and R1040, and is rectified by Q1003 (**IMZ2A**). The resulting DC is fed back through Q1002 (**UMX1**) to the drive amplifier Q1014 and final amplifier Q1008, for control of the power output.

The microprocessor selects "High" or "Low" power levels.

Transmit Inhibit

When the transmit PLL is unlocked, pin 7 of PLL IC Q1005 (**MB15A01PFV**) goes to a logic "Low." The resulting DC unlock control voltage is passed to pin 10 of the microprocessor Q1047. While the transmit PLL is unlocked, pin 31 of Q1047 remains high, which then turns off Q1012 (**CPH6102**) and the Automatic Power Controller Q1002 (**UMX1**) to disable the supply voltage to the drive amplifier Q1014, Q1020 and final amplifier Q1008, thereby disabling the transmitter.

Spurious Suppression

Generation of spurious products by the transmitter is minimized by the fundamental carrier frequency being equal to final transmitting frequency, modulated directly in the transmit VCO. Additional harmonic suppression is provided by a low-pass filter consisting of coils L1002, and L1003 plus capacitors C1001, C1002, C1021, C1023, C1024, C1025, C1026, and C1027, resulting in more than 60 dB of harmonic suppression prior to delivery of the RF signal to the antenna.

PLL Frequency Synthesizer

The PLL circuitry on the Main Unit consists of VCO Q1030 (**2SK210GR**), Q1033 (**2SC5226**), VCO buffer Q1026 (**2SC5005**), and PLL subsystem IC Q1051 (**MB15A01PFV1**), which contains a reference divider, serial-to-parallel data latch, programmable divider, phase comparator and charge pump, and TCXO unit X1002 (**TTS05VS**) which yields frequency stability of ±2.5ppm @ -22°F to +140°F (-30°C to +60°C).

While receiving, VCO Q1030 oscillates between 196.85 and 224.85 MHz according to the transceiver version and the programmed receiving frequency. The VCO output is buffered by Q1026, then applied to the prescaler section of Q1051. There the VCO signal is divided by 64 or 65, according to a control signal from the data latch section of Q1051, before being sent to the programmable divider section of Q1051.

The data latch section of Q1051 also receives serial dividing data from the microprocessor Q1047, which causes the pre-divided VCO signal to be further divided in the programmable divider section, depending upon the desired receive frequency, so as to produce a 2.5 kHz or 3.125 kHz derivative of the current VCO frequency.

Meanwhile, the reference divider section of Q1051 divides the 16.80 MHz crystal reference from the reference oscillator Q1051, by 6720 (or 5376) to produce the 2.5 kHz (or 3.125 kHz) loop references (respectively).

The 2.5 kHz (or 3.125 kHz) signal from the programmable divider (derived from the VCO) and that derived from the reference oscillator are applied to the phase detector section of Q1051, which produces a pulsed output with pulse duration depending on the phase difference between the input signals.

This pulse train is filtered to DC and returned to varactors D1021, D1022, D1023, and D1024 (all **HVD372B**). Changes in the level of the DC voltage are applied to the varactors, affecting the reference in the tank circuit of the VCO according to the phase difference between the signals derived from the VCO and the crystal reference oscillator.

The VCO is thus phase-locked to the crystal reference oscillator. The output of the VCO Q1030 (**25K210GR**), after buffering by Q1026, is applied to the first mixer as described previously.

For transmission, the VCO Q1033 (**2SC5226**) oscillates between 146.00 and 174.00 MHz according to the model version and programmed transmit frequency. The remainder of the PLL circuitry is shared with the receiver. However, the dividing data from the microprocessor is such that the VCO frequency is at the actual transmit frequency (rather than offset for IFs, as in the receiving case). Also, the VCO is modulated by the speech audio applied to D1017 (**HVC350B**), as described previously.

Receive and transmit buses select which VCO is made active, using Q1028, Q1031, Q1032 (all **DTC144EE**).

Miscellaneous Circuits

Push-To-Talk Transmit Activation

The PTT switch on the microphone is connected to pin 22 of microprocessor Q1047, so that when the PTT switch is closed, pin 27 of Q1047 goes high. This signal disables the receiver by disabling the 5 V supply bus at Q1022 (**DTB123EK**) to the front-end, FM IF subsystem IC Q1048 and the receiver VCO circuitry.

At the same time, Q1013 (**UMX1N**) and Q1012 (**CPH6102**) activate the transmit 5V supply line to enable the transmitter.

Introduction

The **VX-410/-420** series has been aligned at the factory for the specified performance across the entire frequency range specified. Realignment should therefore not be necessary except in the event of a component failure. All component replacement and service should be performed only by an authorized Vertex Standard representative, or the warranty policy may be voided.

The following procedures cover the sometimes critical and tedious adjustments that are not normally required once the transceiver has left the factory. However, if damage occurs and some parts are replaced, realignment may be required. If a sudden problem occurs during normal operation, it is likely due to component failure; realignment should not be done until after the faulty component has been replaced.

We recommend that servicing be performed only by authorized Vertex Standard service technicians who are experienced with the circuitry and fully equipped for repair and alignment. Therefore, if a fault is suspected, contact the dealer from whom the transceiver was purchased for instructions regarding repair. Authorized Vertex Standard service technicians realign all circuits and make complete performance checks to ensure compliance with factory specifications after replacing any faulty components. Those who do undertake any of the following alignments are cautioned to proceed at their own risk. Problems caused by unauthorized attempts at realignment are not covered by the warranty policy. Also, Vertex Standard must reserve the right to change circuits and alignment procedures in the interest of improved performance, without notifying owners. Under no circumstances should any alignment be attempted unless the normal function and operation of the transceiver are clearly understood, the cause of the malfunction has been clearly pinpointed and any faulty components replaced, and the need for realignment determined to be absolutely necessary. The following test equipment (and thorough familiarity with its correct use) is necessary for complete realignment. Correction of problems caused by misalignment resulting from use of improper test equipment is not covered under the warranty policy. While most steps do not require all of the equipment listed, the interactions of some adjustments may require that more complex adjustments be performed afterwards. Do not attempt to perform only a single step unless it is clearly isolated electrically from all other steps. Have all test equipment ready before beginning, and follow all of the steps in a section in the order presented.

Required Test Equipment

- Avionics Radio Tester with calibrated output level at 500 MHz
- □ In-line Wattmeter with 5% accuracy at 500 MHz
- □ 50-ohm, 10-W RF Dummy Load
- □ Regulated DC Power Supply (standard 7.5V DC, 2A)
- □ Frequency Counter: ±0.2 ppm accuracy at 500 MHz
- □ AF Signal Generator
- □ AC Voltmeter
- DC Voltmeter
- VHF Sampling Coupler
- IBM® PC/Compatible Computer with Microsoft® Windows® 95 or later operating system
- □ Vertex Standard CE64 Alignment program
- Vertex Standard FIF-12 USB Programming Interface and CT-104A, CT-106, or CT-171 PC Programming Cable
- Vertex Standard FRB-6 Tuning Interface Box and CT-160 Connection Cable

Alignment Preparation & Precautions

A 50-ohm RF Dummy load and in-line wattmeter must be connected to the main antenna jack in all procedures that call for transmission, except where specified otherwise. Correct alignment is not possible with an antenna.

After completing one step, read the following step to determine whether the same test equipment will be required. If not, remove the test equipment (except dummy load and wattmeter, if connected) before proceeding.

Correct alignment requires that the ambient temperature be the same as that of the transceiver and test equipment, and that this temperature be held constant between 20° and 30°C. When the transceiver is brought into the shop from hot or cold air, it should be allowed time to come to room temperature before alignment.

Whenever possible, alignments should be made with oscillator shields and circuit boards firmly affixed in place. Also, the test equipment must be thoroughly warmed up before beginning.

Note:Signal levels in dB referred to in this procedure are based on 0 dB μ = 0.5 μ V (closed circuit).

Important Note -

When connecting the **CT-160** plug into the **MIC/SP** jack of the **VX-410/-420** series, you must remove the plastic cap and its mounting screws prior to programming. Please remember to re-attach the cap and screws when the programming is complete.

Test Setup

Set up the test equipment as shown below for transceiver alignment, and apply 7.5V DC power to the transceiver.

The transceiver must be programmed for use in the intended system before alignment is attempted. The RF parameters are loaded from the file during the alignment process.

In order to facilitate alignment over the complete operating rang of the equipment, it is recommended that the channel data in the transceiver be preset as per the chart below.

The alignment tool outline

Installation of the Alignment tool

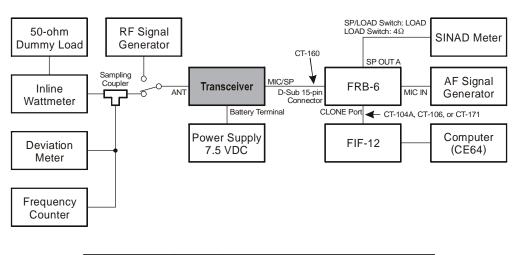
The "alignment mode" is a software-based protocol, accessed by an "Alignment Mode" command from the computer while switching the transceiver on. It is operated by the alignment tool automatically. During use of the alignment mode, normal operation is suspended. The alignment tool program provides all needed operation capability.

Alignment Sequence

Although the data displayed on the computer's screen during alignment is temporary data, it is important you follow the basic alignment sequence precisely, so that the displayed data and the data loaded into the transceiver are identical.

Basic Alignment Sequence

- 1. Enter the alignment mode
- 2. Upload data from transceiver
- 3. Align data
- 4. Download data to transceiver



CHANNEL	Түре	FREQUENCY	CTCSS TONE	DCS CODE
Low Band Edge	А	134.000 MHz		
(Channel 1)	С	146.000 MHz		
Band Center	А	147.000 MHz	254.1 Hz	
(Channel 2)	С	160.000 MHz	204.1 ПZ	
High Band Edge	А	160.000 MHz		325
(Channel 3)	С	174.000 MHz		525

PLL VCV (Varactor Control Voltage)

- □ Connect the DC voltmeter between **TP1045** on the Main Unit and ground.
- □ Set the transceiver to CH 3 (high band edge), and adjust L1036 on the Main Unit for 3.7 V ± 0.1 V on the DC voltmeter.
- □ Set the transceiver to CH 1 (low band edge), and confirm the low-end VCV is more than 1.2 V while receiving.
- □ Set the transceiver to CH 3 (high band edge), and adjust **L1038** on the Main Unit for 3.6 V ± 0.1 V while transmitting.
- □ Set the transceiver to CH 1 (low band edge), and confirm the low-end VCV is more than 1.1 V while transmitting.

Reference Frequency

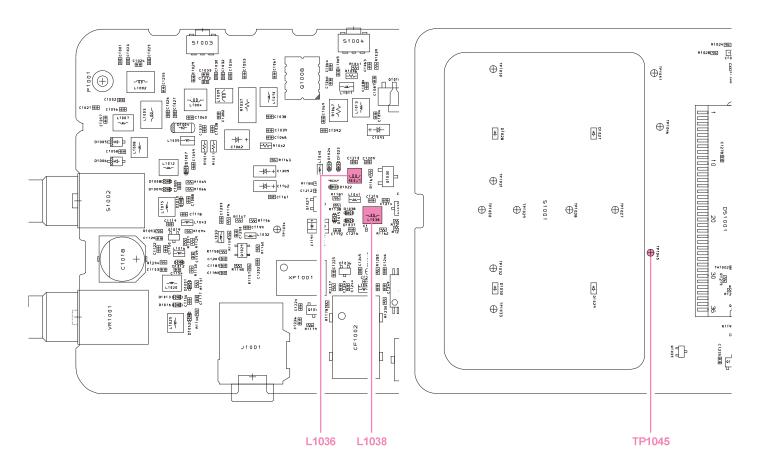
- Set the transceiver to CH 2 (band center) via the "CH" box on the "Alignment" window previously.
- □ To adjustment, click the left mouse button on the "**RF Frequency**" box, then press the [**ENTER**] key to open the pop-up window.
- □ Use the [←] or [→] arrow keys so that the frequency counter displays the band center frequency (±100 Hz) for the version being aligned.
- \square Press the "OK" box to lock in the new data.

Transmitter Output Power

- □ Set the transceiver to CH 2 (band center).
- □ To adjustment, click the left mouse button on the "**RF Power {High}**" or "**RF Power {Low}**" box, then press the [**ENTER**] key to open the pop-up window.
- □ Use the [←] or [→] arrow keys so that the power meter reading is 5.0 W (± 0.1 W) (for "RF Power High") or 1.0 W (± 0.1 W) (for "RF Power Low"). Confirm that the current consumption is 2.3 A or lower (for "RF Power High") or 1.0 A or lower (for "RF Power Low").
- □ Press the "OK" box to lock in the new data.

MAX Deviation

- □ Set the transceiver to CH 2 (band center).
- □ Inject a 1 kHz tone at −17 dBm to the **MIC** jack.
- To adjustment, click the left mouse button on the "MAX Deviation" box, then press the [ENTER] key to open the pop-up window.
- □ Use the [←] or [→] arrow keys so that the deviation meter reading is ±4.3 kHz (±0.1 kHz) (for 25 kHz steps) or ±2.1 kHz (±0.1 kHz) (for 12.5 kHz steps) deviation.
- $\hfill\square$ Press the "OK" box to lock in the new data.



CTCSS Deviation

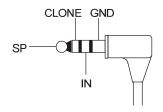
- □ Set the transceiver to CH 2 (band center).
- □ To adjustment, click the left mouse button on the "CTCSS Modulation" box, then press the [ENTER] key to open the pop-up window.
- □ Use the [←] or [→] arrow keys so that the deviation meter reading is ±0.8 kHz (±0.1 kHz) (for 25 kHz steps) or ±0.5 kHz (±0.1 kHz) (for 12.5 kHz steps) deviation.
- □ Press the "OK" box to lock in the new data.

MAX Ref Deviation

- □ Set the transceiver to CH 2 (band center).
- □ To adjustment, click the left mouse button on the "MAX **Ref Deviation**" box, then press the [ENTER] key to open the pop-up window.
- □ Use the [←] or [→] arrow keys so that the deviation meter reading is ±0.5 kHz (±0.1 kHz) (for 25 kHz steps) or ±0.4 kHz (±0.1 kHz) (for 12.5 kHz steps) deviation.
- □ Set the transceiver to CH 2 (band center), then key the transmitter, and confirm that the deviation is ±0.6 kHz ~ ±0.9 kHz (for 25 kHz steps) or ±0.35 kHz ~ ±0.6 kHz (for 12.5 kHz steps).
- □ Press the "OK" box to lock in the new data.

STD Deviation

- □ Set the transceiver to CH 2 (band center).
- □ Inject a 1 kHz tone at −37 dBm to the **MIC** jack.



- To adjustment, click the left mouse button on the "MIC Sensitivity" box, then press the [ENTER] key to open the pop-up window.
- □ Use the [←] or [→] arrow keys so that the deviation meter reading is ±3.0 kHz (±0.1 kHz) (for 25 kHz steps) or ±1.5 kHz (±0.1 kHz) (for 12.5 kHz steps) deviation.
- \Box Press the "OK" box to lock in the new data.

DTMF Deviation

- □ Set the transceiver to CH 2 (band center).
- □ To adjustment, click the left mouse button on the "DTMF Deviation" box, then press the [ENTER] key to open the pop-up window.
- □ Use the [←] or [→] arrow keys so that the deviation meter reading is ±3.0 kHz (±0.1 kHz) (for 25 kHz steps) or ±1.5 kHz (±0.1 kHz) (for 12.5 kHz steps) deviation.
- \square Press the "OK" box to lock in the new data.

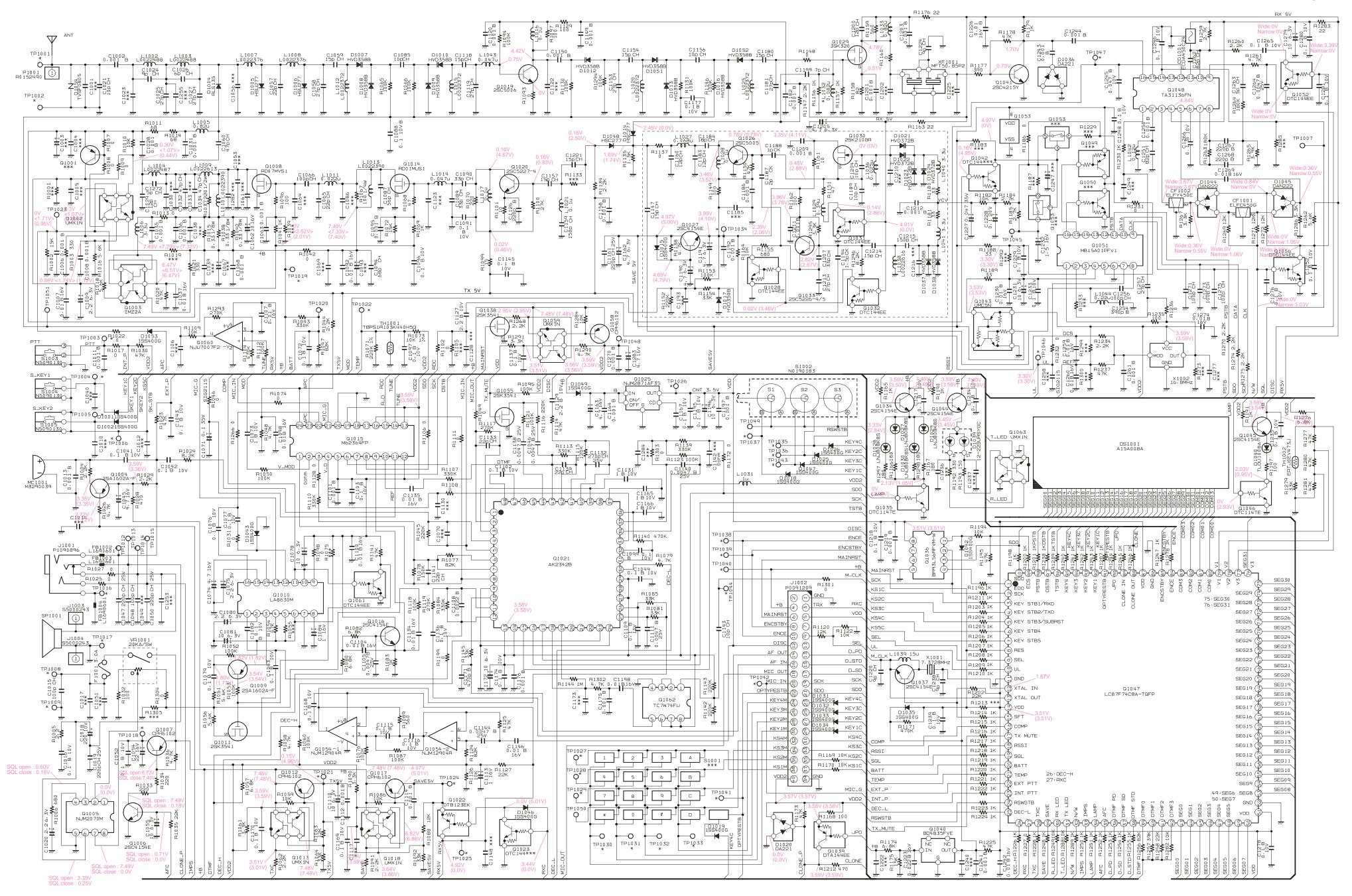
DCS Deviation

- □ Set the transceiver to CH 2 (band center).
- To adjustment, click the left mouse button on the "DCS Modulation" box, then press the [ENTER] key to open the pop-up window.
- □ Use the [←] or [→] arrow keys so that the deviation meter reading is ±0.7 kHz (±0.1 kHz) (for 25 kHz steps) or ±0.45 kHz (±0.1 kHz) (for 12.5 kHz steps) deviation.
- \square Press the "OK" box to lock in the new data.

Sensitivity

- □ Set the transceiver to CH 3 (high band edge).
- □ Tune the RF signal generator to the same frequency as the transceiver's, then set the generator output level to 40 dBµ with ±3.0 kHz deviation @ 1 kHz tone modulation.
- To adjustment, click the left mouse button on the "RX Sensitivity" box, then press the [ENTER] key to open the pop-up window.
- □ Use the [\leftarrow] or [\rightarrow] arrow keys to tune for best sensitivity; ultimately, the radio should be aligned so that the RF signal generator output level is -6 dBµ EMF (0.25 µV) or less for 12 dB SINAD.
- □ Press the "OK" box to lock in the new data.

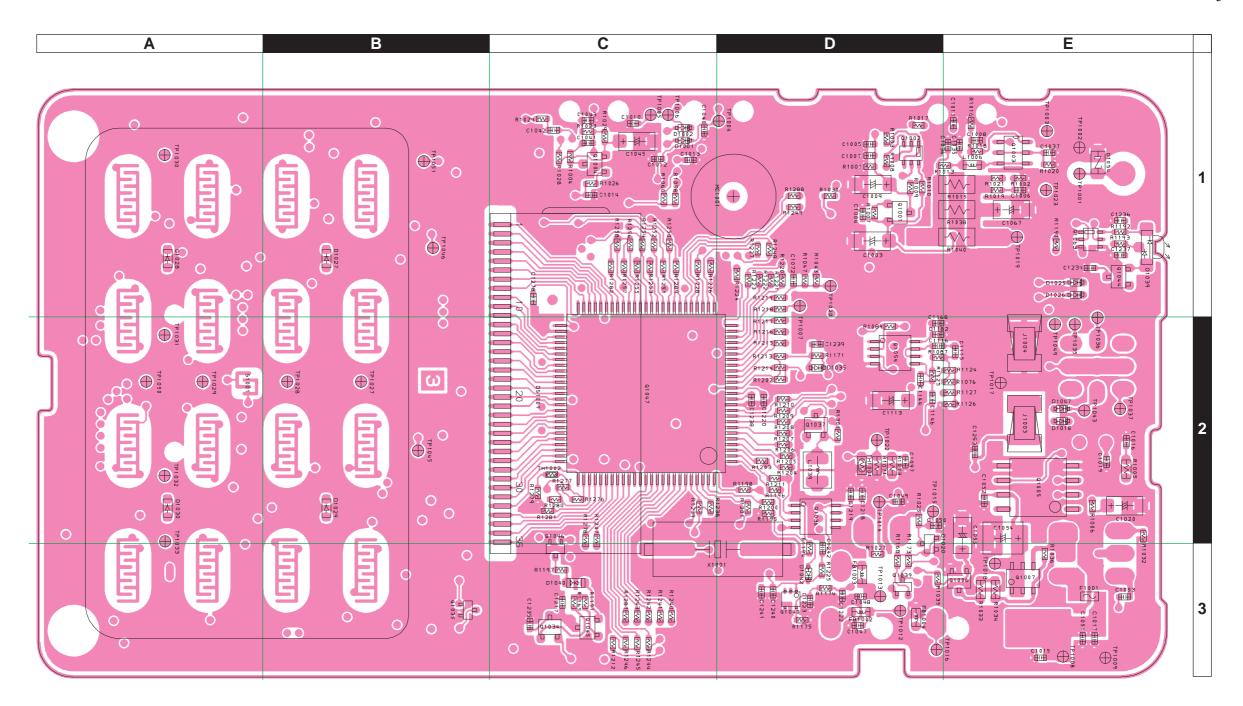
This completes the internal alignment routine. To save all settings and exit, press the "OK" box.



MAIN Unit (Lot. 1~2)

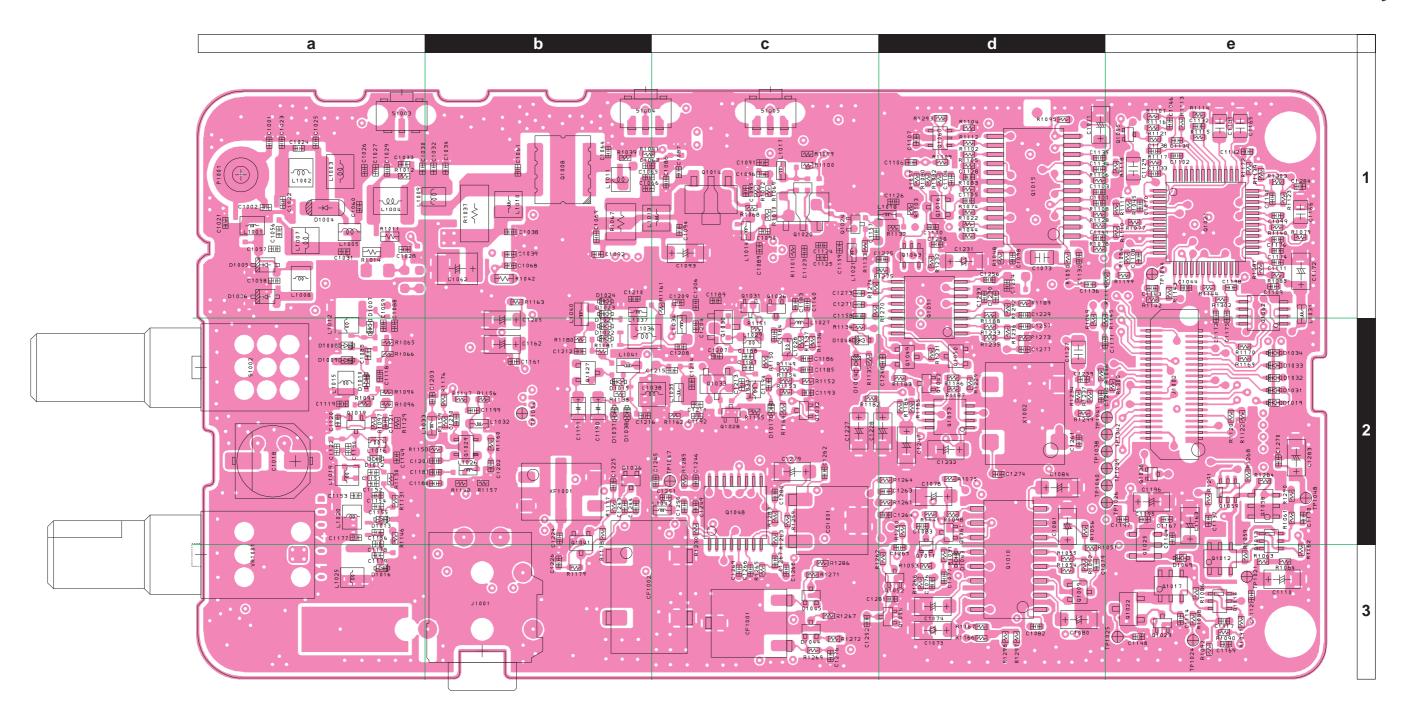
Circuit Diagram

RX : XX TX : (XX) TX HIGH : <XX> TX LOW : {XX} LED on : [XX]



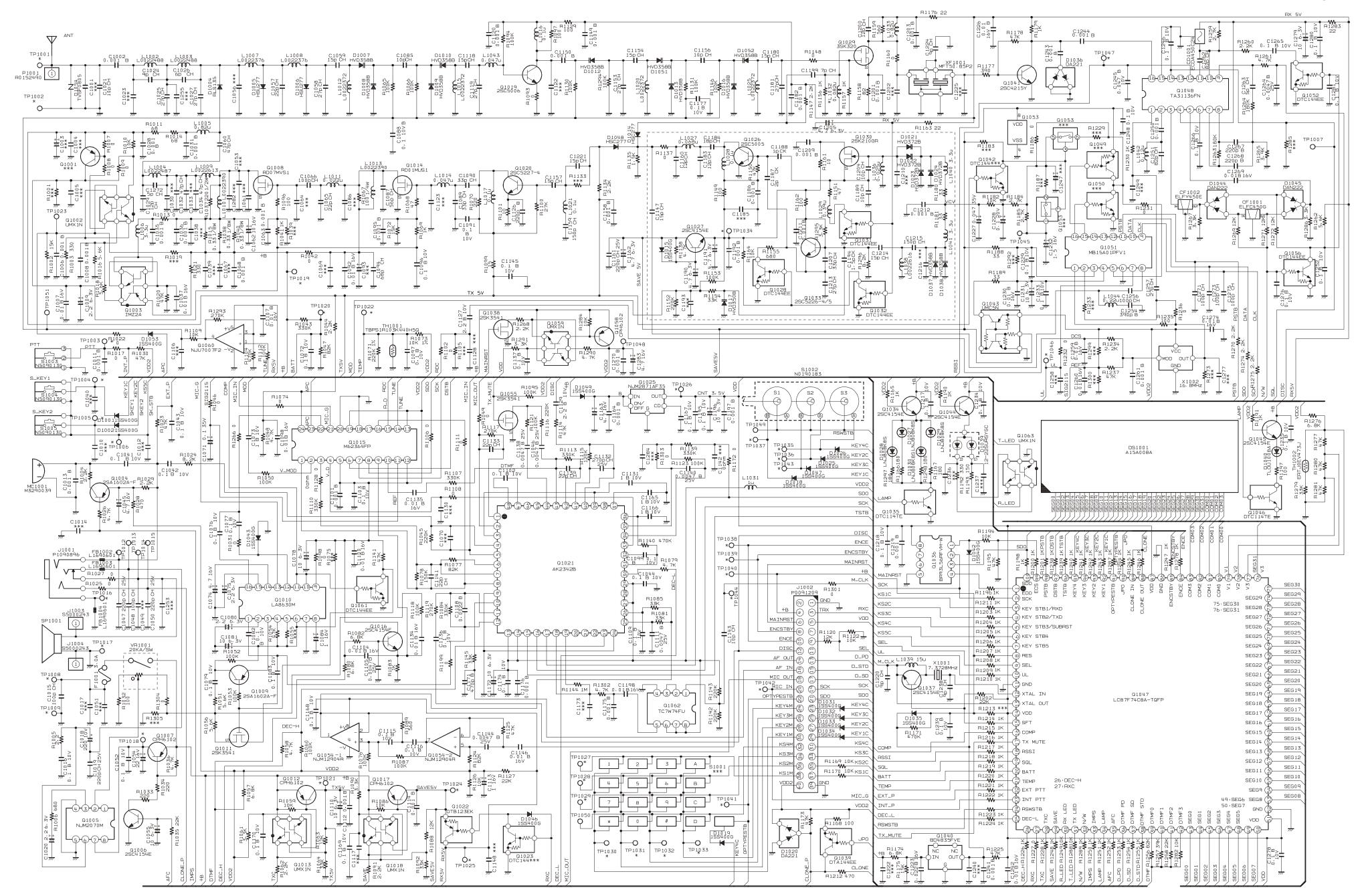
MAIN Unit (Lot. 1~2)

Parts Layout (Side A)



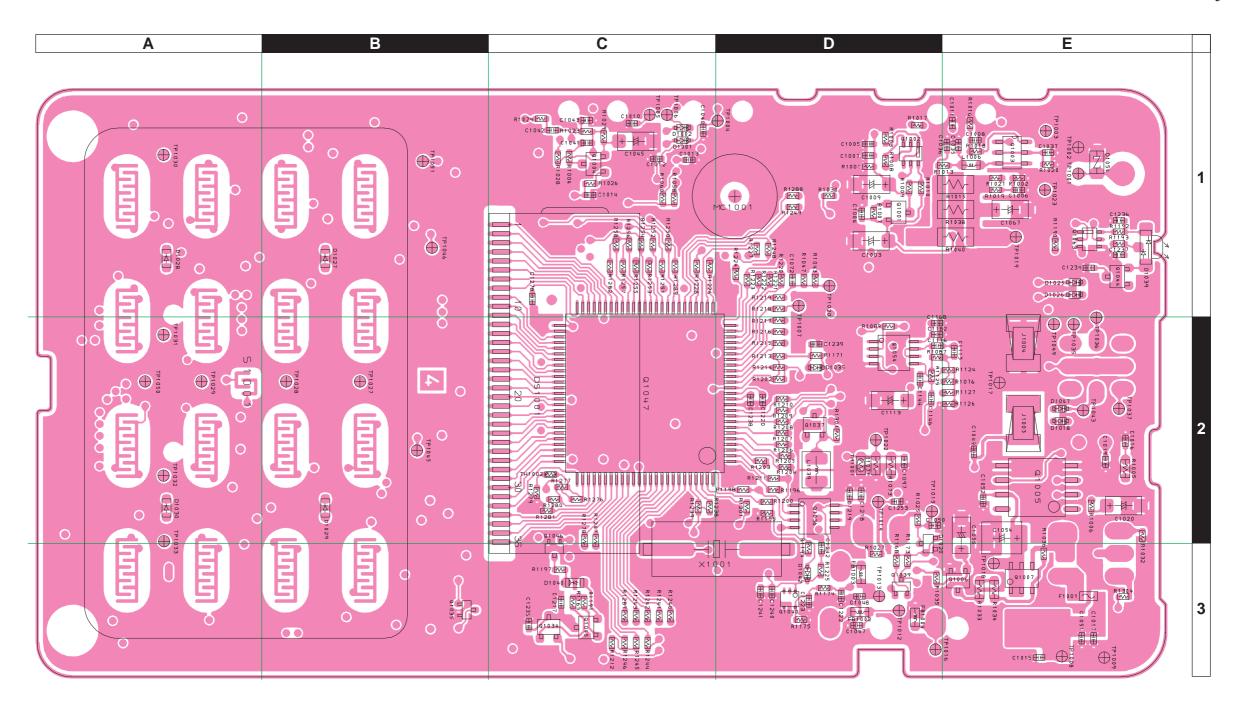
MAIN Unit (Lot. 1~2)

Parts Layout (Side B)



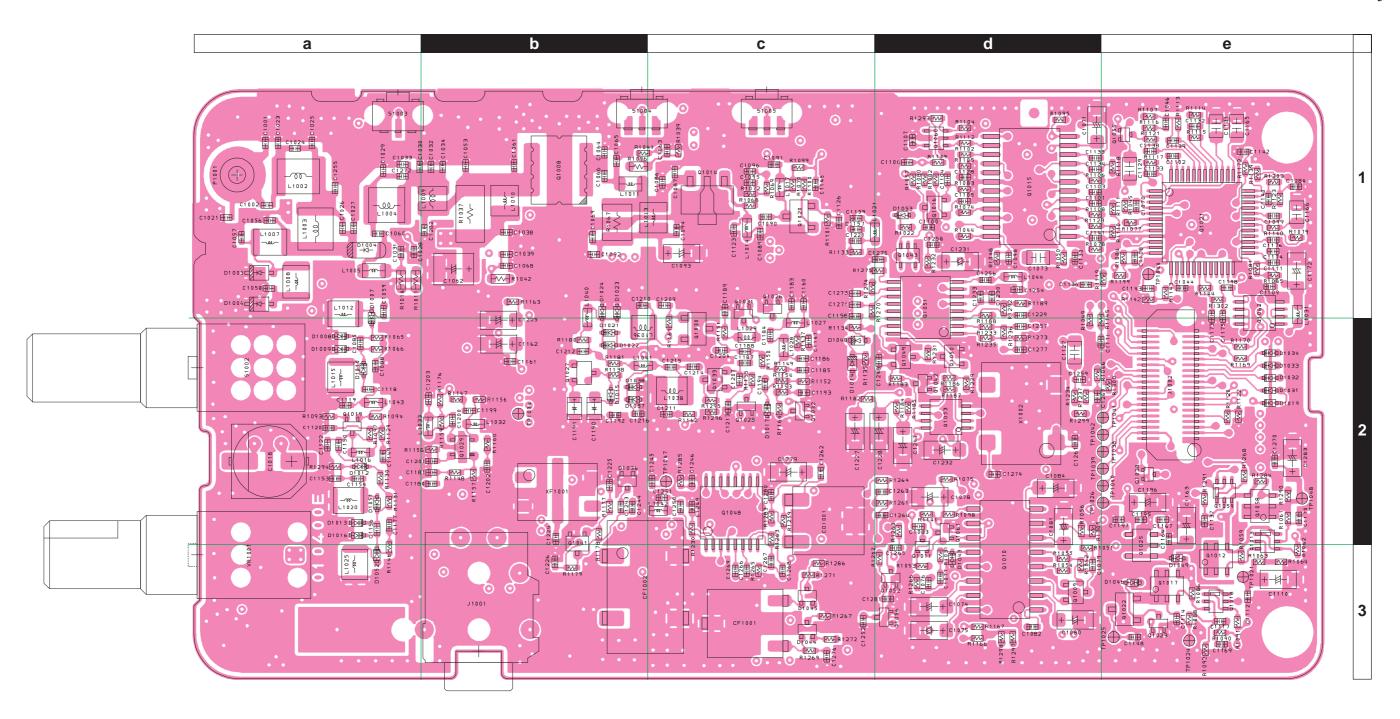
MAIN Unit (Lot. 3~8)

Circuit Diagram



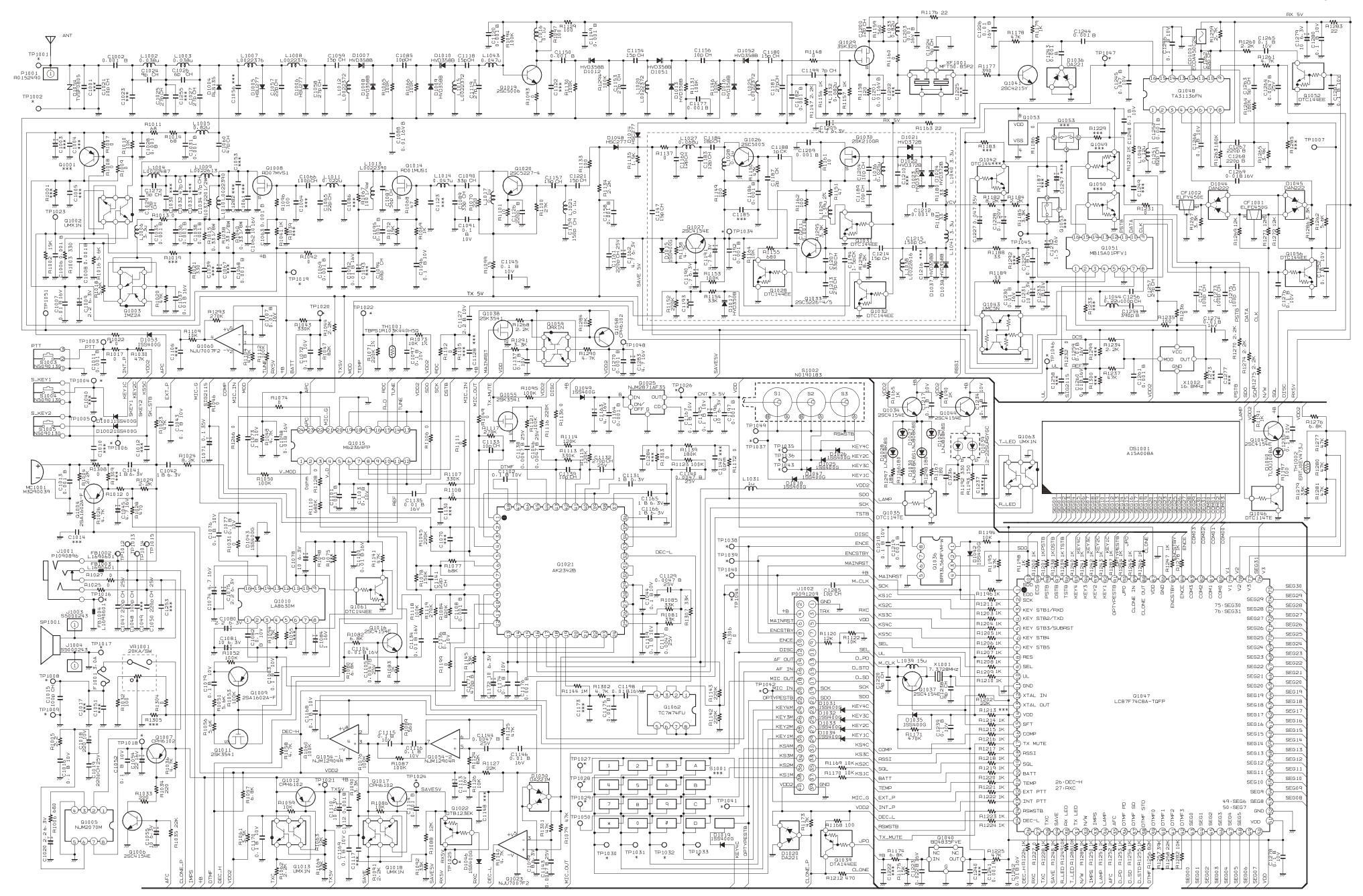
MAIN Unit (Lot. 3~8)

Parts Layout (Side A)



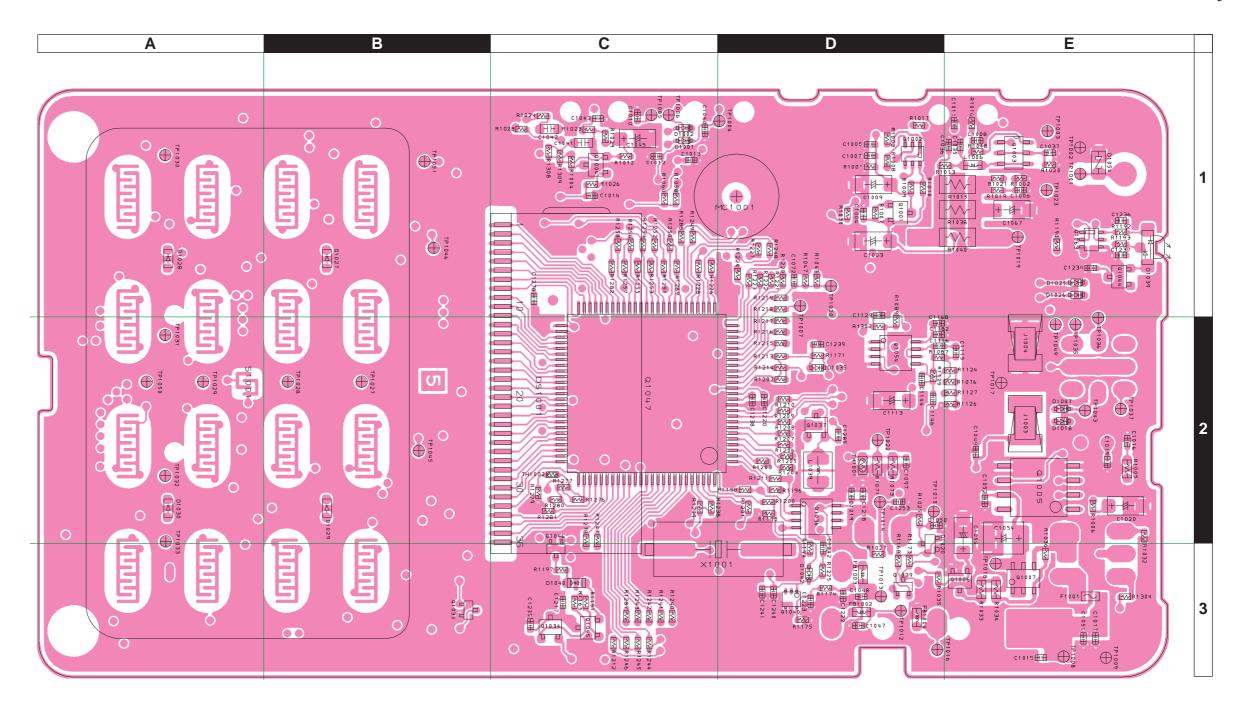
MAIN Unit (Lot. 3~8)

Parts Layout (Side B)



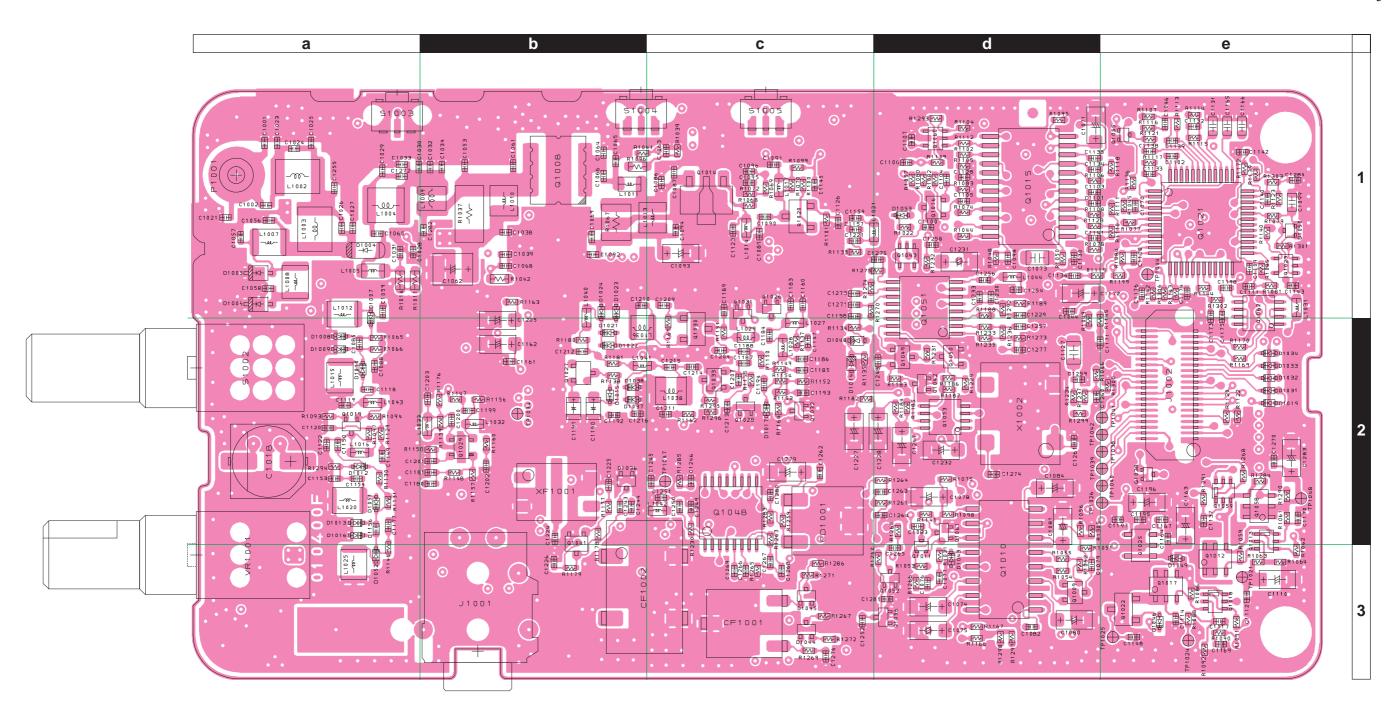
MAIN Unit (Lot. 9~28)

Circuit Diagram



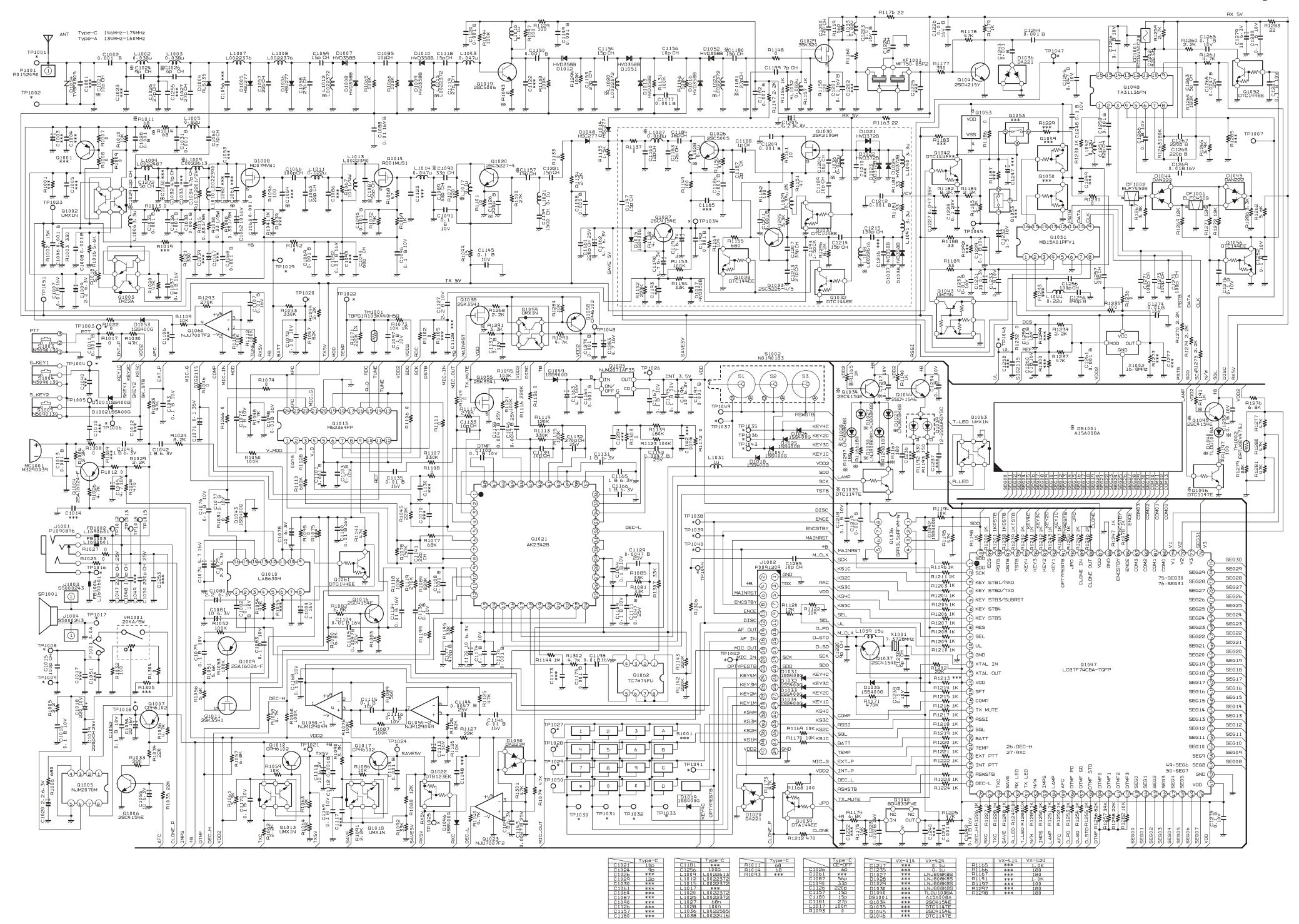
MAIN Unit (Lot. 9~28)

Parts Layout (Side A)



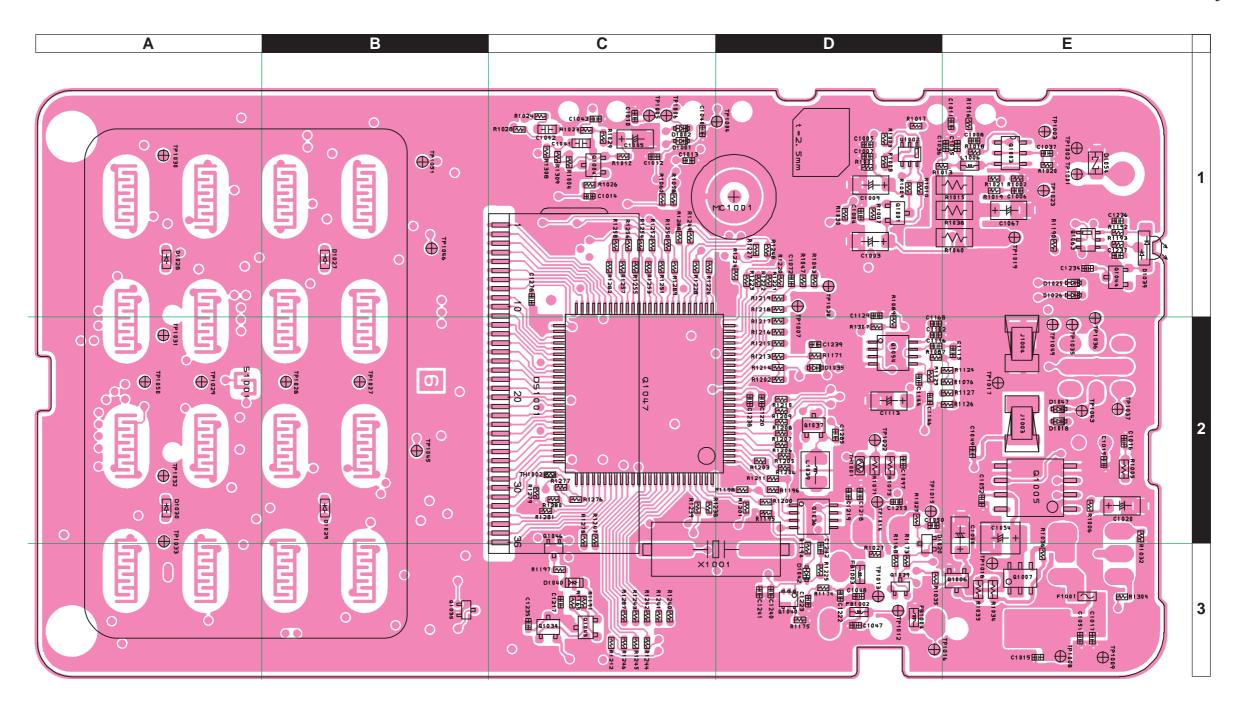
MAIN Unit (Lot. 9~28)

Parts Layout (Side B)



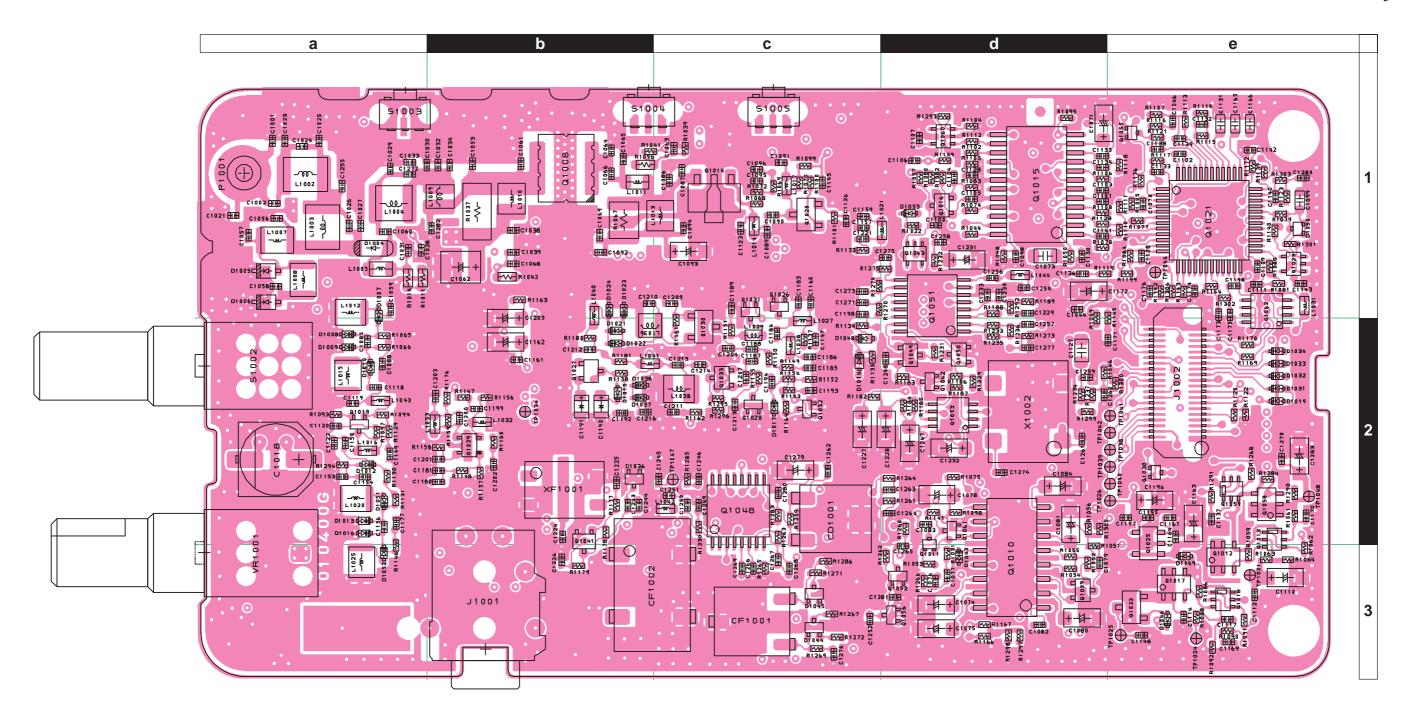
MAIN Unit (Lot. 29~35)

Circuit Diagram



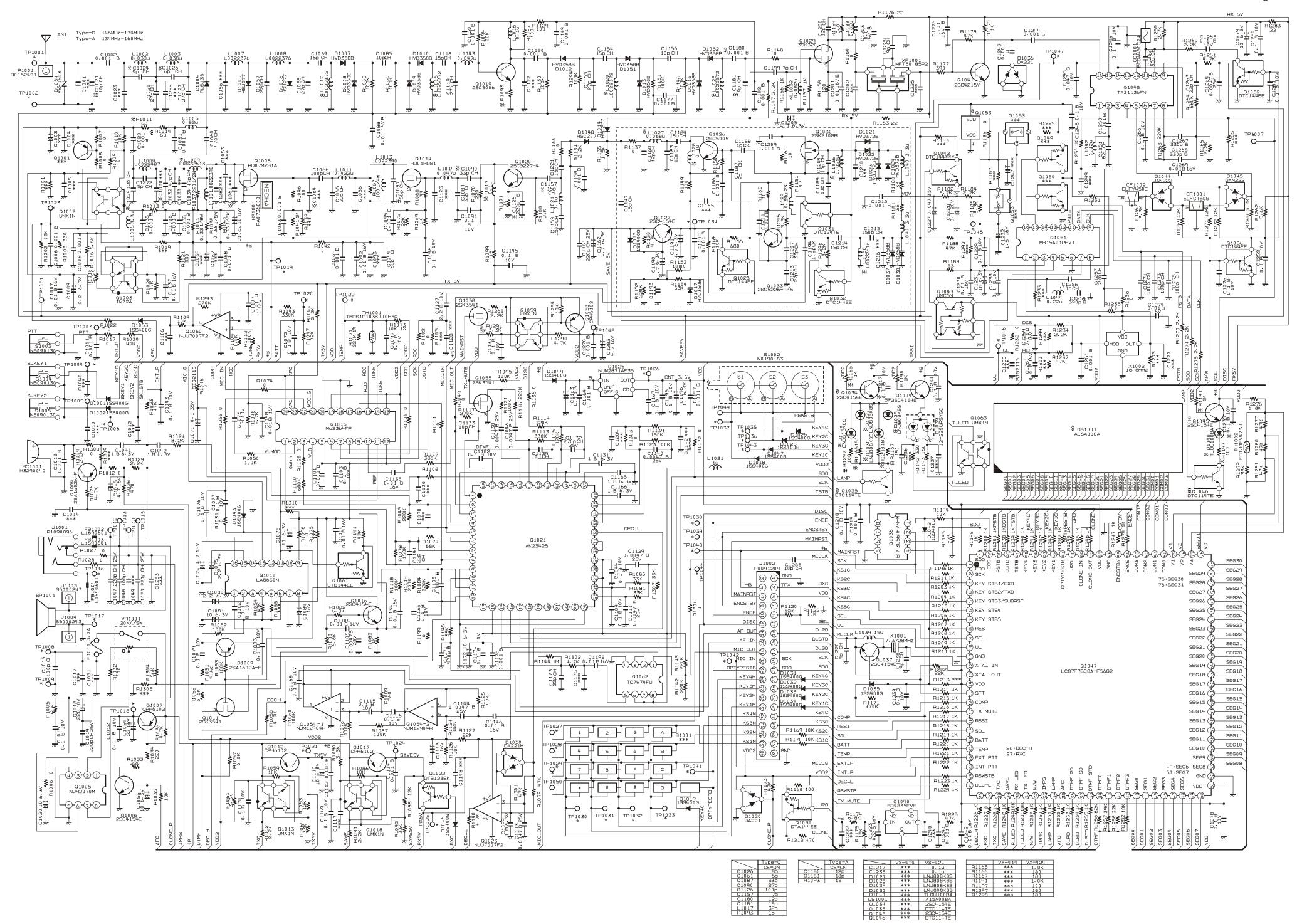
MAIN Unit (Lot. 29~35)

Parts Layout (Side A)



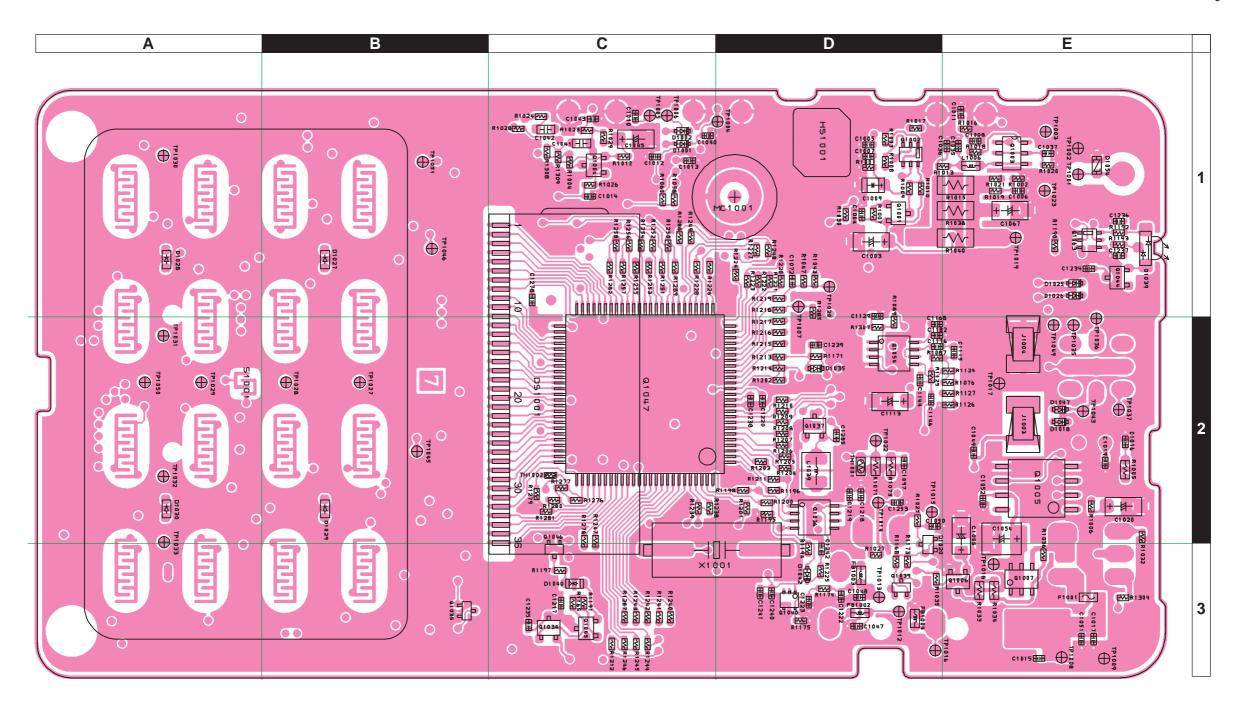
MAIN Unit (Lot. 29~35)

Parts Layout (Side B)



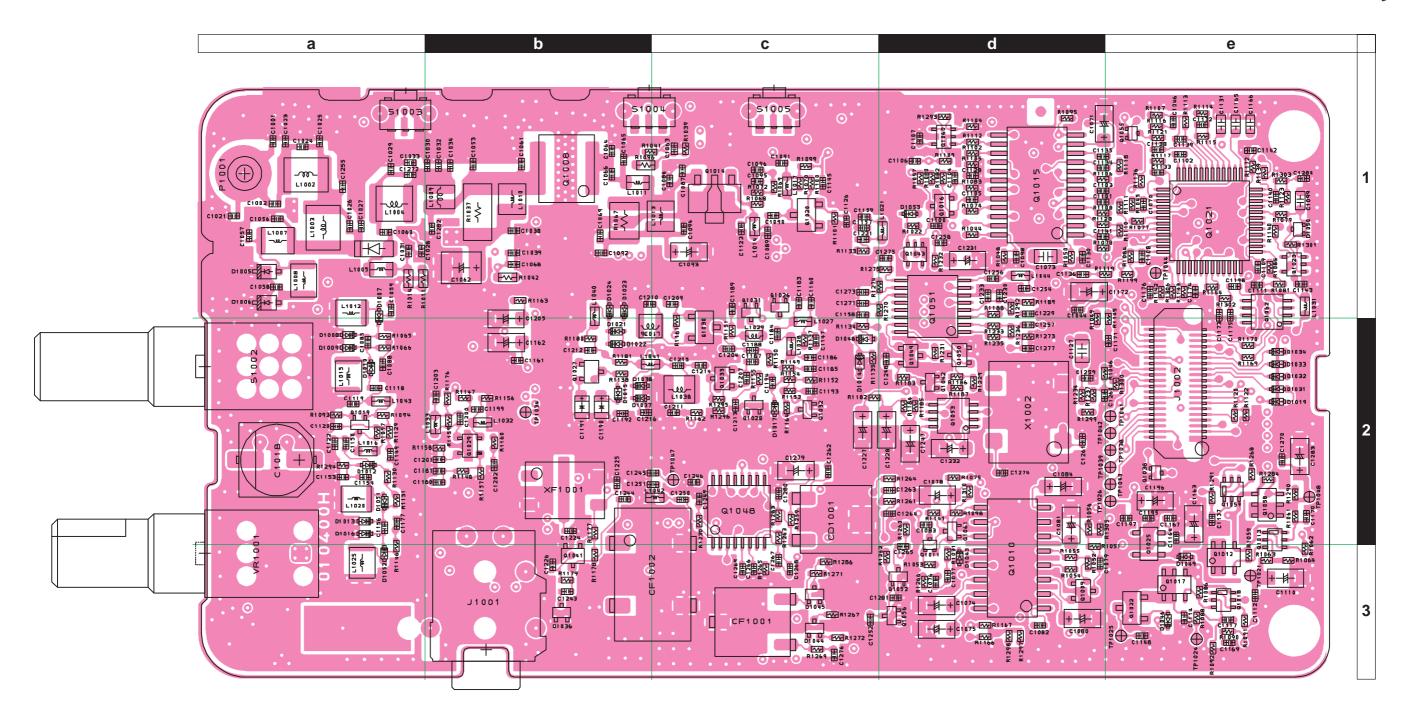
MAIN Unit (Lot. 36~53)

Circuit Diagram



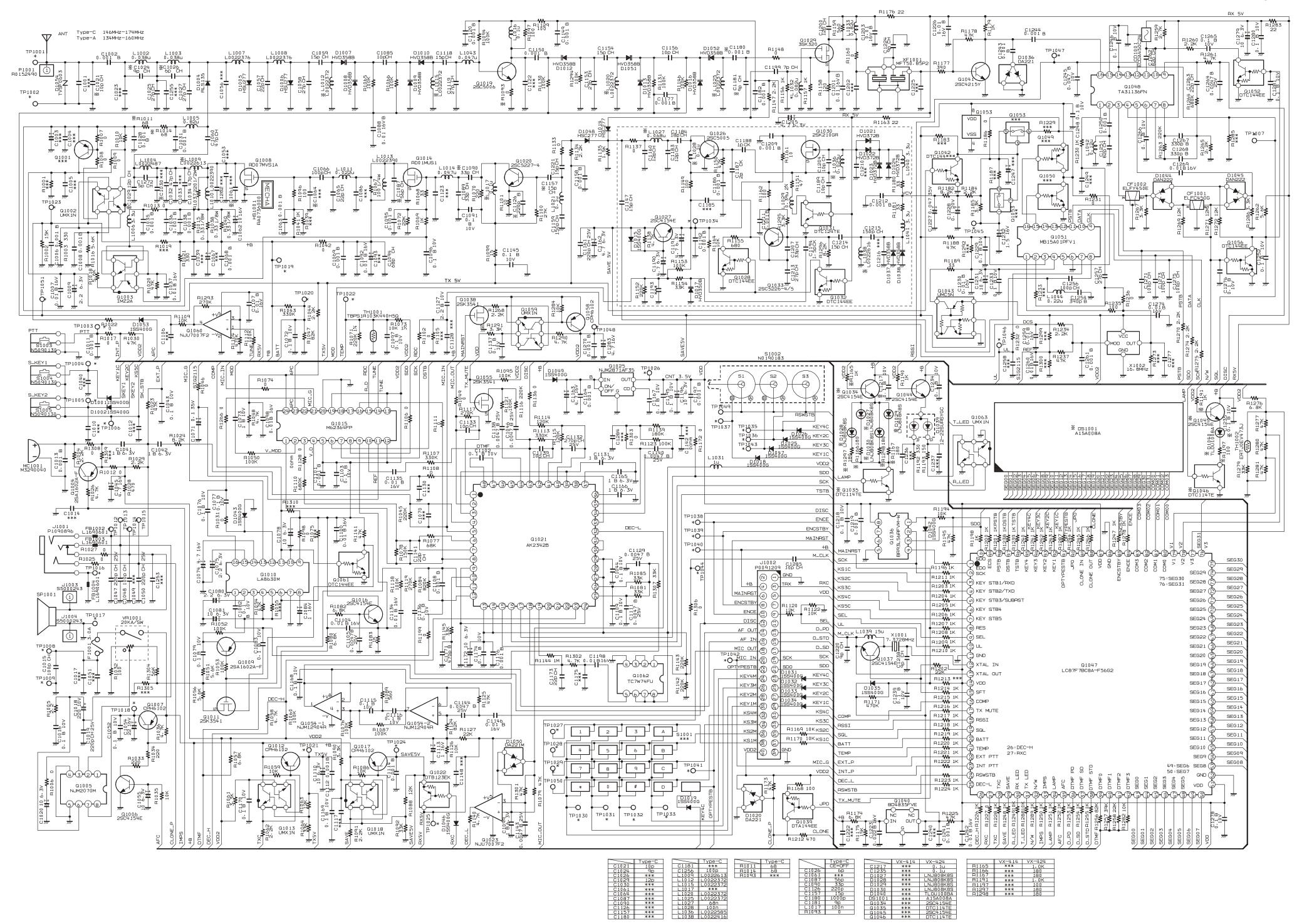
MAIN Unit (Lot. 36~53)

Parts Layout (Side A)



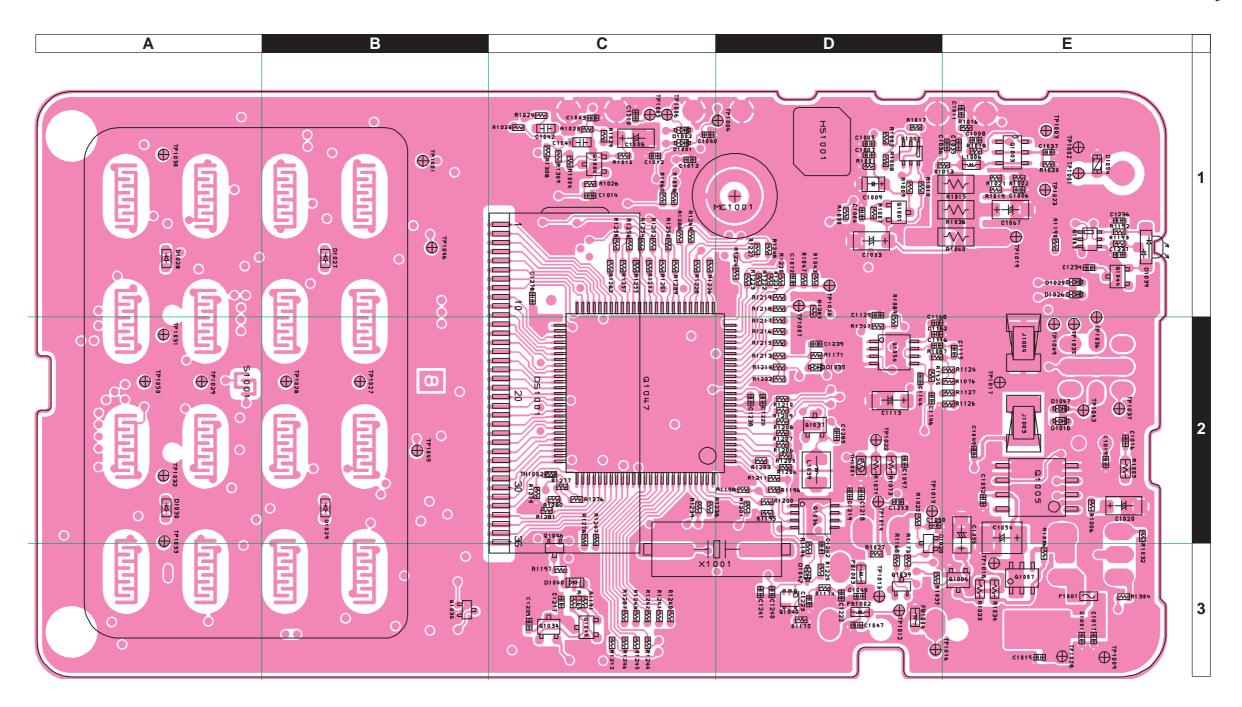
MAIN Unit (Lot. 36~53)

Parts Layout (Side B)



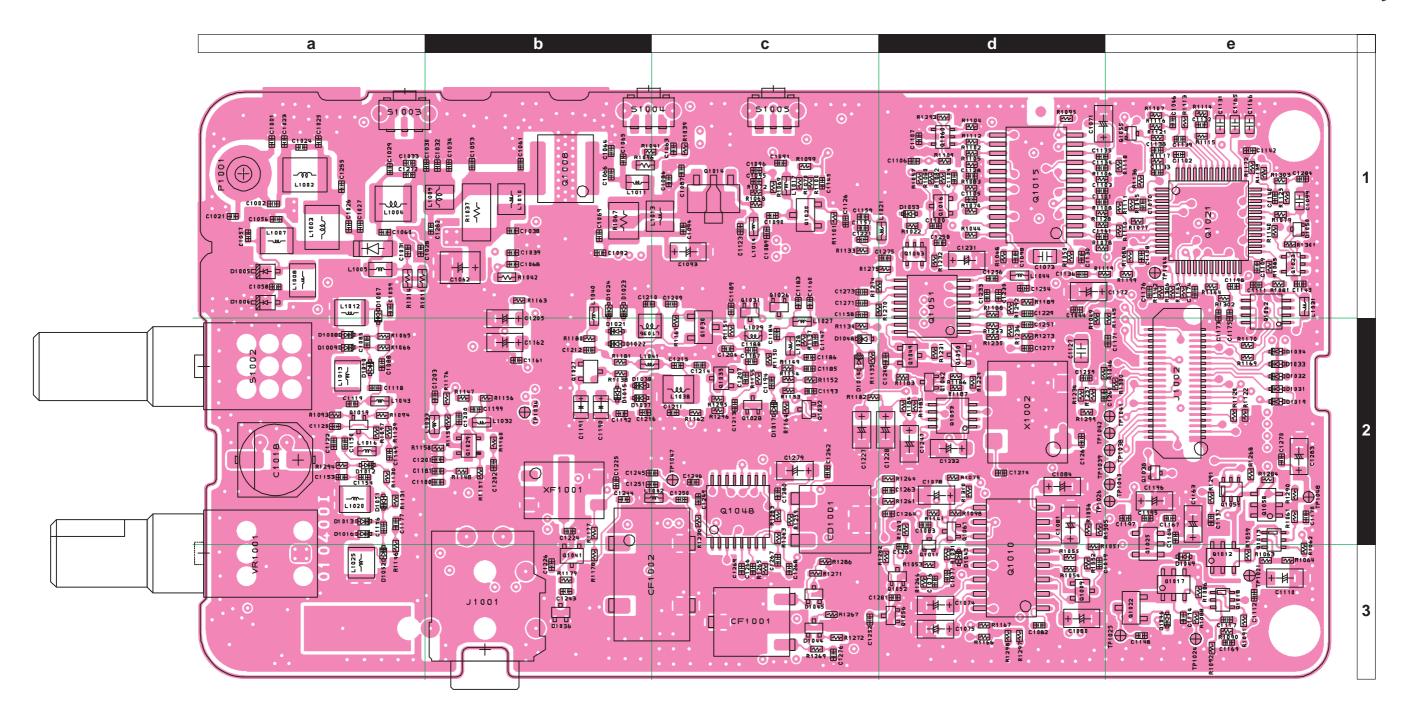
MAIN Unit (Lot. 54~)

Circuit Diagram



MAIN Unit (Lot. 54~)

Parts Layout (Side A)

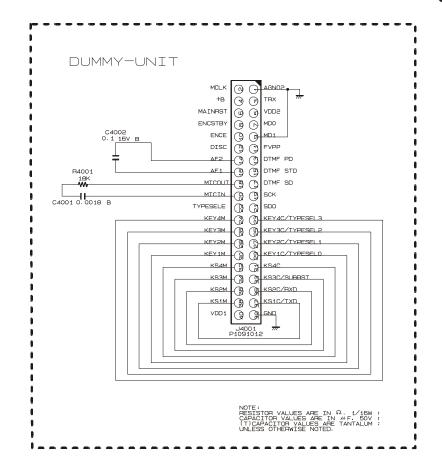


MAIN Unit (Lot. 54~)

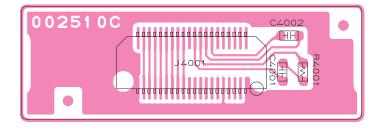
Parts Layout (Side B)

DUMMY Unit

Circuit Diagram

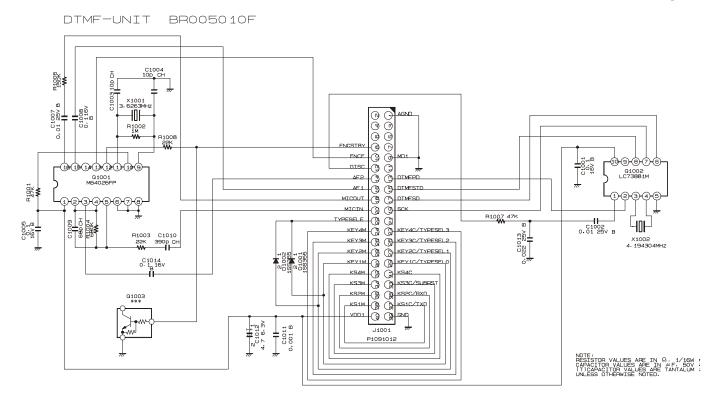


Parts Layout

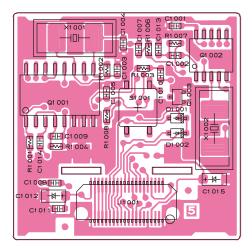


FVP-25 Encryption / DTMF Pager Unit (Option)

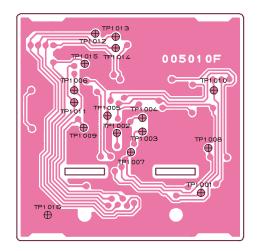
Circuit Diagram



Parts Layout

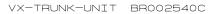


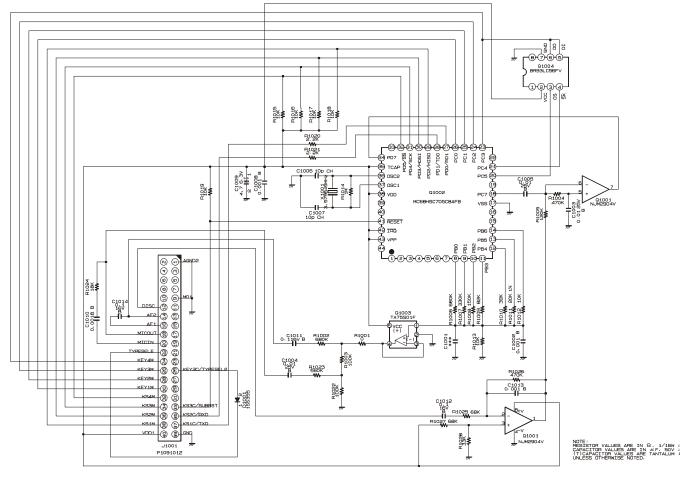
Side A



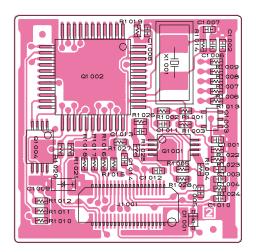
Side B

Circuit Diagram

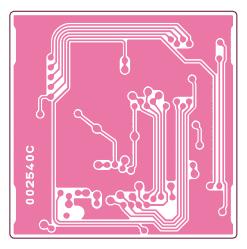




Parts Layout



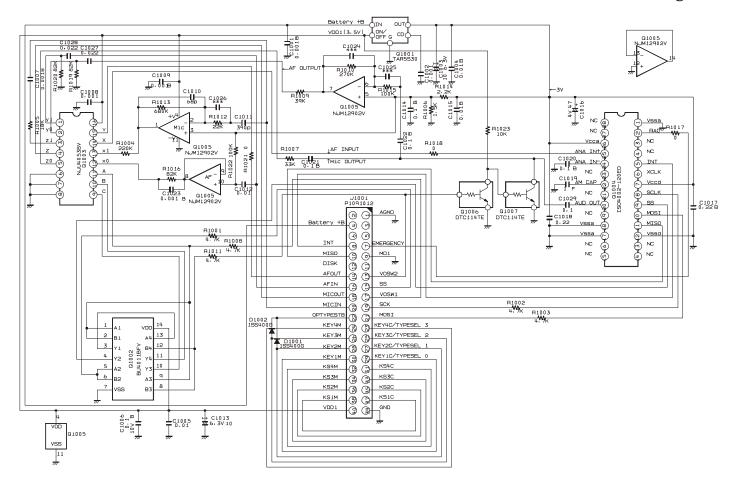
Side A



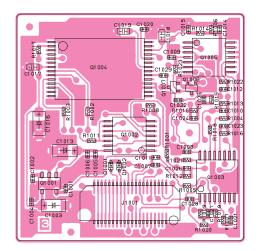
Side B

DVS-5 Voice Storage Unit (Option)

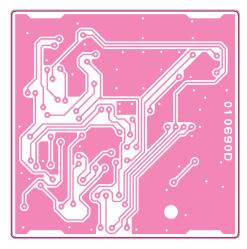
Circuit Diagram



Parts Layout



Side A



Side B



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